

Macroscopic Emitters

on

Deployable Structures

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Supported by AFOSR

Outline

- 1) Antenna array on deployable structures
 - a. Far-field formula
 - b. Constructive interference
 - c. Applications with illustrative examples

- 2) General time-harmonic solution of Maxwell's equation
 - a. Trapezoidal rule
 - b. Constructive/destructive interference
 - c. Structure determination

Background: Maxwell's equation

$$\begin{aligned}\nabla \cdot \mathbf{E} &= \frac{\rho}{\epsilon_0} & \nabla \times \mathbf{E} &= -\frac{\partial \mathbf{B}}{\partial t} \\ \nabla \cdot \mathbf{B} &= 0 & \nabla \times \mathbf{B} &= \mu_0 \mathbf{J} + \mu_0 \epsilon_0 \frac{\partial \mathbf{E}}{\partial t}\end{aligned}$$

Two relevant results in monochromatic regime $\mathbf{E}(\mathbf{x}, t) = e^{-i\omega t} \mathbf{E}_0(\mathbf{x})$

Field due to oscillating charge at mean position $\bar{\mathbf{y}}$ (Antenna radiation)

$$\mathbf{E}_0(\mathbf{x}) = k_{el} \frac{1}{|\mathbf{x} - \bar{\mathbf{y}}|} \left(\mathbf{I} - \frac{\mathbf{x} - \bar{\mathbf{y}}}{|\mathbf{x} - \bar{\mathbf{y}}|} \otimes \frac{\mathbf{x} - \bar{\mathbf{y}}}{|\mathbf{x} - \bar{\mathbf{y}}|} \right) \ddot{\mathbf{y}} e^{i(\omega/c)|\mathbf{x} - \bar{\mathbf{y}}|}$$

Radiation in free space

$$\begin{aligned}\Delta \mathbf{E}_0 &= -\frac{\omega^2}{c^2} \mathbf{E}_0 \\ \nabla \cdot \mathbf{E}_0 &= 0\end{aligned}$$

Antenna model using oscillating charge

- For single antenna (modelled by an oscillating charge),

$$\mathbf{E}_0(\mathbf{x}) = k_{el} \frac{1}{|\mathbf{x} - \bar{\mathbf{y}}|} \left(\mathbf{I} - \frac{\mathbf{x} - \bar{\mathbf{y}}}{|\mathbf{x} - \bar{\mathbf{y}}|} \otimes \frac{\mathbf{x} - \bar{\mathbf{y}}}{|\mathbf{x} - \bar{\mathbf{y}}|} \right) \bar{\mathbf{y}} e^{i(\omega/c)|\mathbf{x} - \bar{\mathbf{y}}|}$$

$\bar{\mathbf{y}}$: Mean position of charge

k_{el} : Electronic constant

$\bar{\mathbf{y}}$: Mean acceleration of charge

ω : Time frequency of the radiation

- For multiple antennas (using superposition),

$$\mathbf{E}_0(\mathbf{x}) = k_{el} \frac{1}{|\mathbf{x} - \mathbf{y}_s|} \left(\mathbf{I} - \frac{\mathbf{x} - \mathbf{y}_s}{|\mathbf{x} - \mathbf{y}_s|} \otimes \frac{\mathbf{x} - \mathbf{y}_s}{|\mathbf{x} - \mathbf{y}_s|} \right) \sum_j \bar{\mathbf{y}}_j e^{i\phi_j} e^{i(\omega/c)|\mathbf{x} - \mathbf{y}_s|}$$

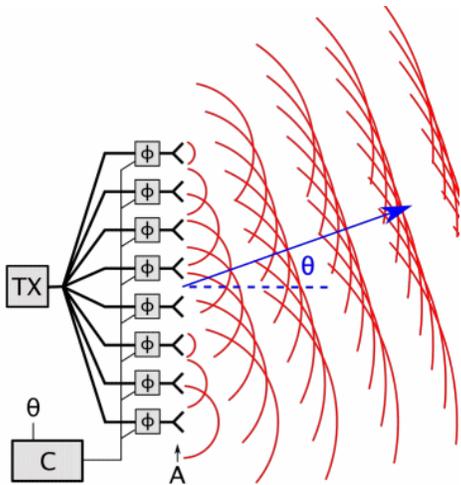
\mathbf{y}_s : Mean position of antenna array

ϕ_j : Relative phase between antenna

Radiation generated by antenna array

Using simple substitution : $\phi_j + (\omega/c) |\mathbf{x} - \mathbf{y}_j| = \varphi_j$

$$\mathbf{E}_0(\mathbf{x}) = k_{el} \frac{1}{|\mathbf{x} - \mathbf{y}_s|} \left(\mathbf{I} - \frac{\mathbf{x} - \mathbf{y}_s}{|\mathbf{x} - \mathbf{y}_s|} \otimes \frac{\mathbf{x} - \mathbf{y}_s}{|\mathbf{x} - \mathbf{y}_s|} \right) \sum_j \bar{\mathbf{y}} e^{i\varphi_j}$$



Relative phase between antennas

parameterized by k_1 and k_2 : $\varphi_j = f_j(k_1, k_2)$

Free parameters
(related to group)

f_j 's are chosen as to get the desired interference pattern in parameter space

Isometry groups

$$g = (\mathbf{Q}|\mathbf{c}), \quad \mathbf{Q} \in O(3), \quad \mathbf{c} \in \mathbb{R}^3$$

$$g(\mathbf{x}) = \mathbf{Q}\mathbf{x} + \mathbf{c}, \quad \mathbf{x} \in \mathbb{R}^3$$

$$g_1 = (\mathbf{R}_1|\mathbf{c}_1) \quad g_2 = (\mathbf{R}_2|\mathbf{c}_2) \quad g_1 g_2 = (\mathbf{R}_1 \mathbf{R}_2 | \mathbf{c}_1 + \mathbf{R}_1 \mathbf{c}_2) \quad id = (\mathbf{I}|\mathbf{0})$$

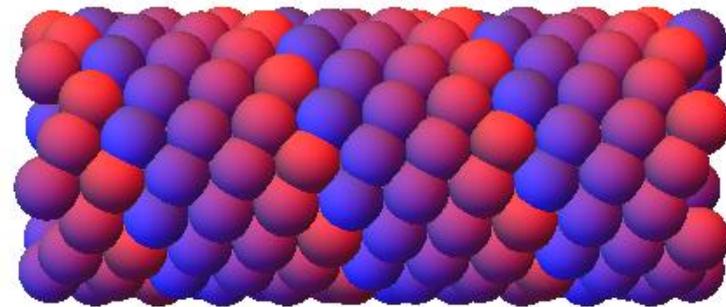
Group product Identity

Helical group (Abelian, two generators)

$$\{h^i g^j : i \in \mathbb{Z}, j = 1, \dots, n\}$$

$$h = (\mathbf{R}_\theta | \tau \mathbf{e} + (\mathbf{R}_\theta - \mathbf{I})\mathbf{x}_0}$$

$$g = (\mathbf{R}_{2\pi/n} | (\mathbf{R}_{2\pi/n} - \mathbf{I})\mathbf{x}_0}$$



Antenna position given by group orbits

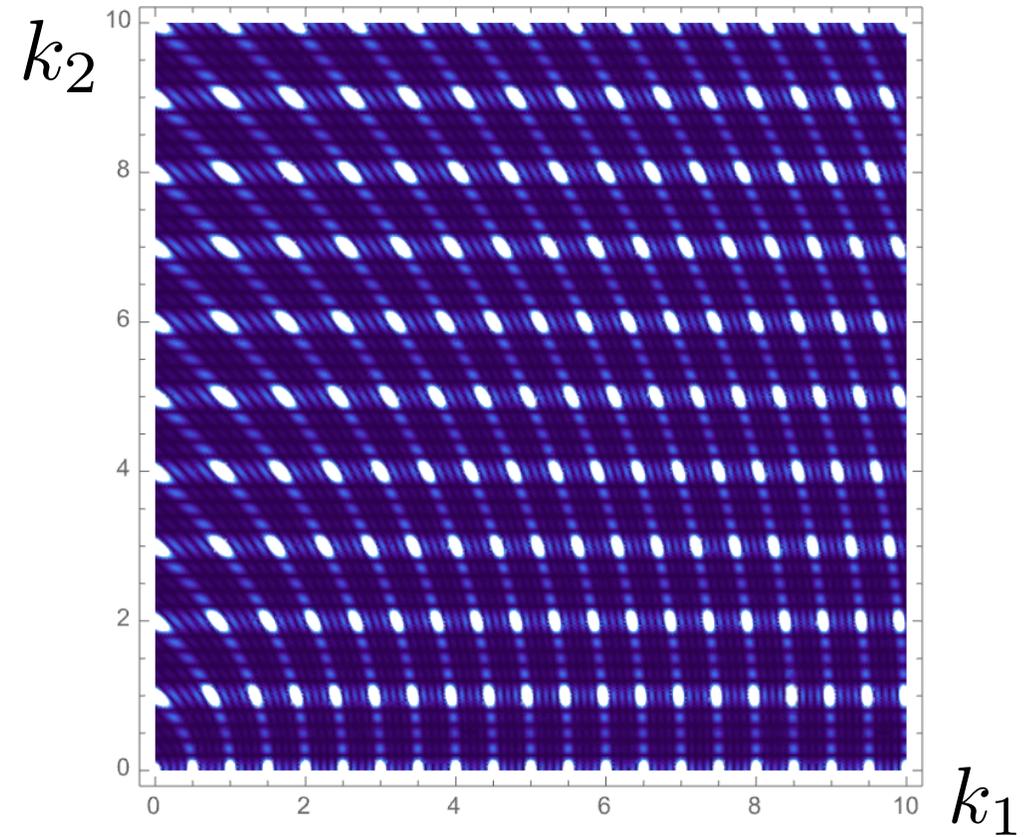
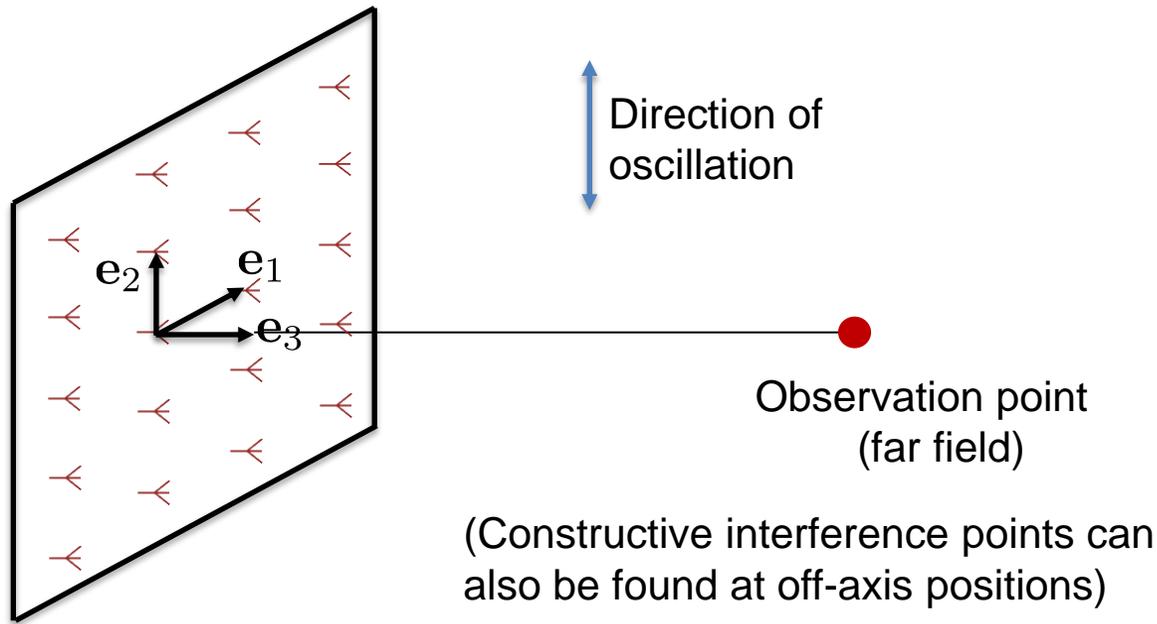
Translation group

$$g_1 = (\mathbf{I} \mid a\mathbf{e}_1)$$

$$g_2 = (\mathbf{I} \mid b\mathbf{e}_2)$$

Position of antenna ℓm $\mathbf{y}_{\ell m} = g_1^\ell g_2^m (\mathbf{y}_0)$

Phase of antenna ℓm $\varphi_{\ell m} = 2\pi \left(\ell k_1 + m k_2 + \ell \sqrt{k_1^2 + k_2^2} \right)$



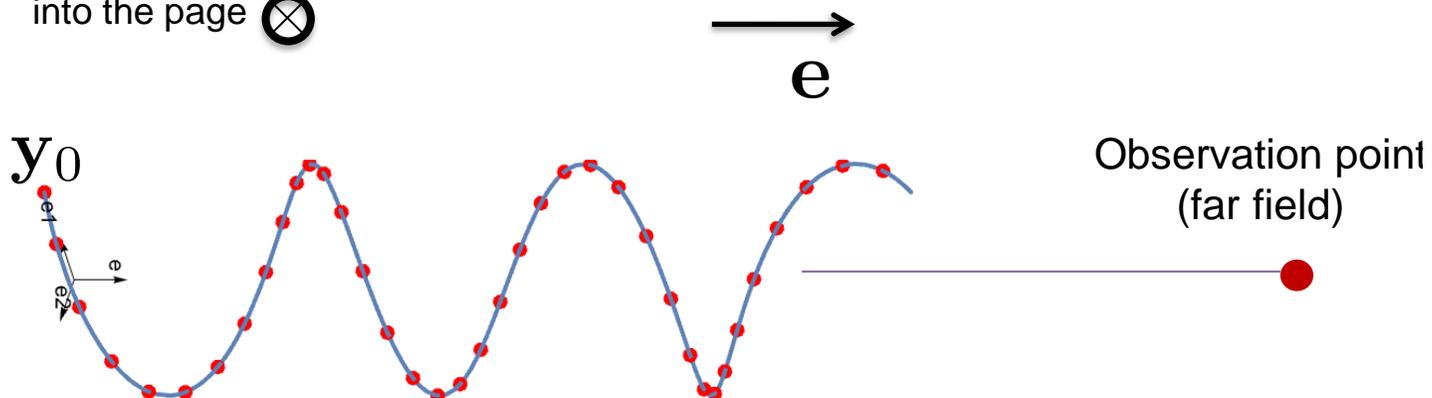
Helical group

$$g = (\mathbf{R}_\theta | \tau \mathbf{e})$$

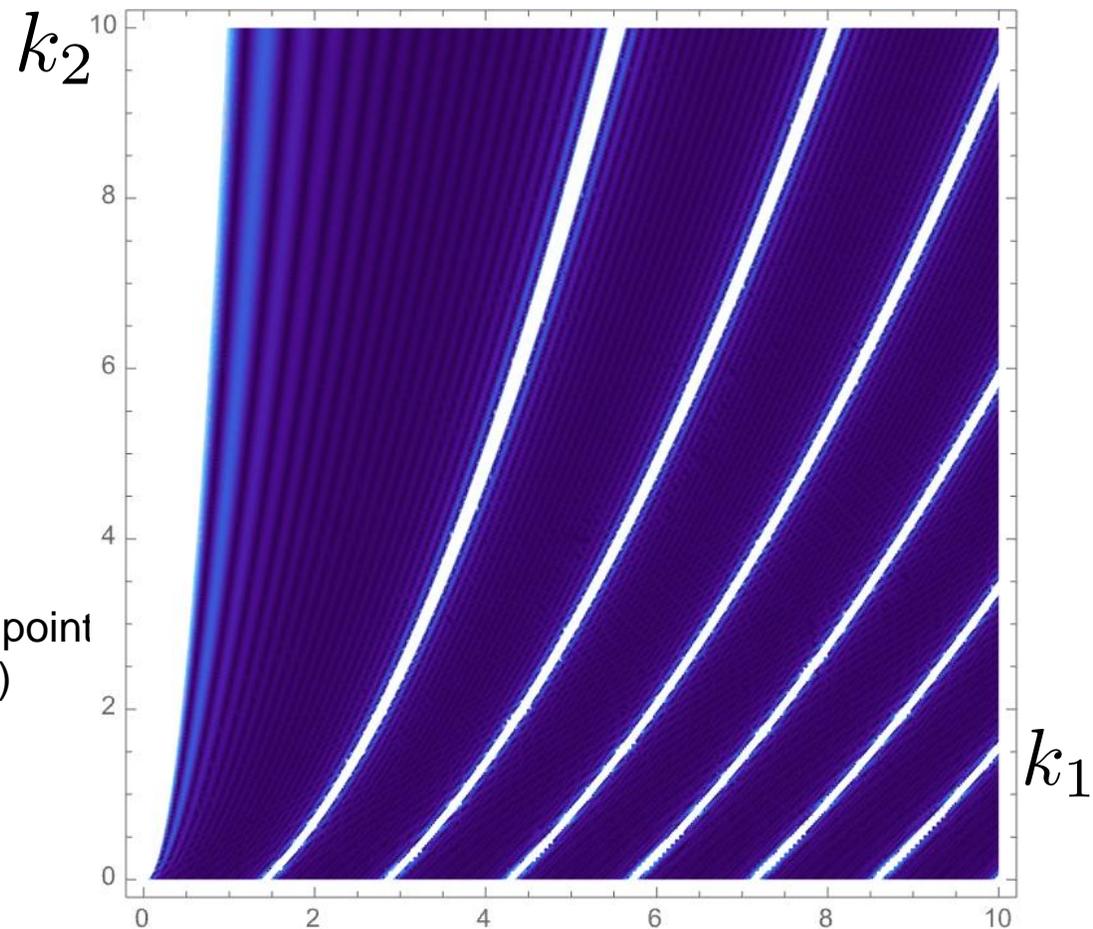
$$\mathbf{y}_j = g^j (\mathbf{y}_0)$$

$$\varphi_j = 2\pi \left(k_1 + j\tau \left(k_2 - \sqrt{k_1^2 + k_2^2} \right) \right)$$

Direction of oscillation:
into the page \otimes



Intensity



Conformal group

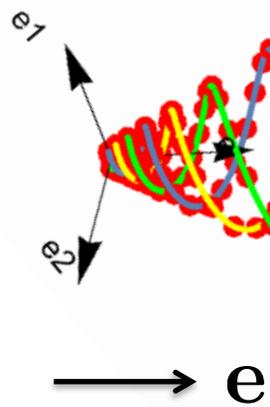
$$g_1 = (\lambda \mathbf{R}_\theta \mid \tau \mathbf{e})$$

$$g_2 = (\mathbf{R}_{2\pi/n} \mid 0)$$

$$\mathbf{y}_{\ell m} = g_1^\ell g_2^m (\mathbf{y}_0)$$

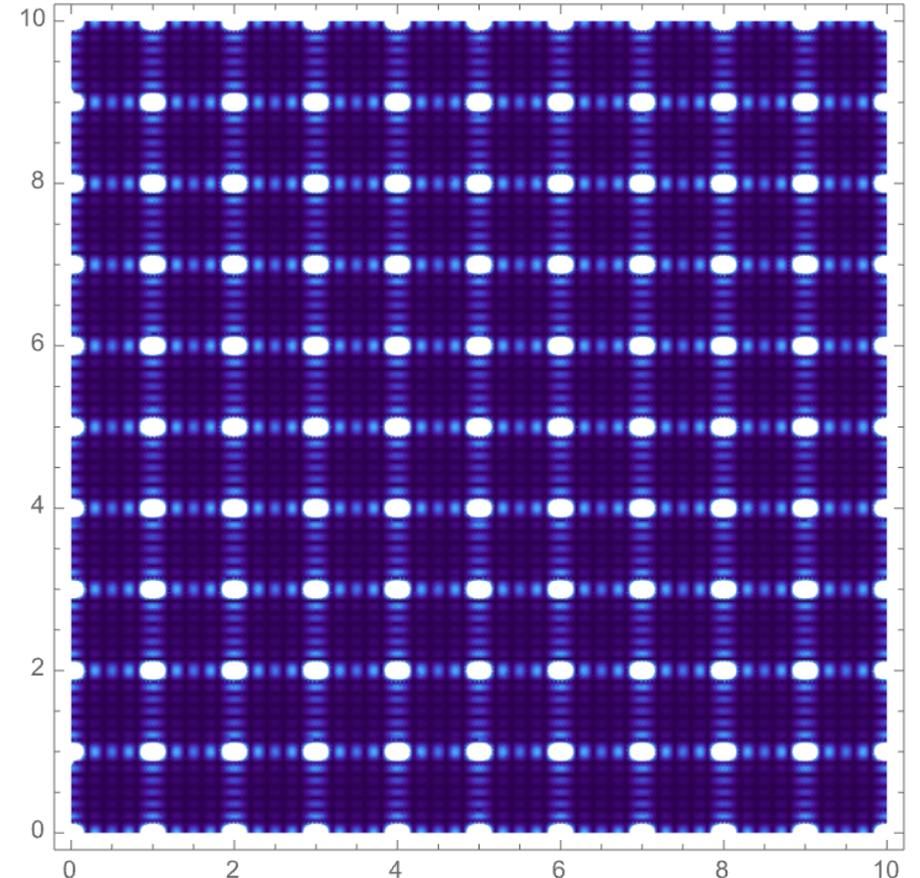
$$\varphi_{\ell m} = 2\pi (\ell k_1 + m k_2)$$

Direction of oscillation:
into the page \otimes



Observation point
(far field)

k_2



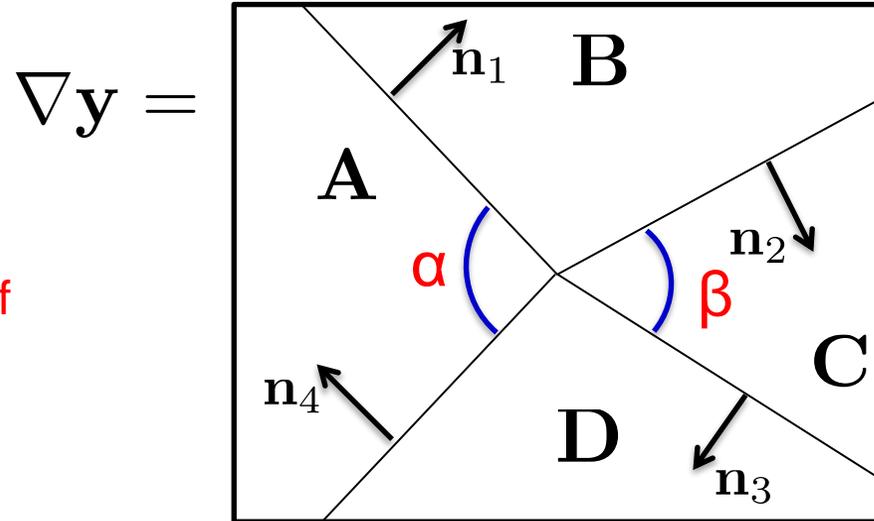
k_1

Summary

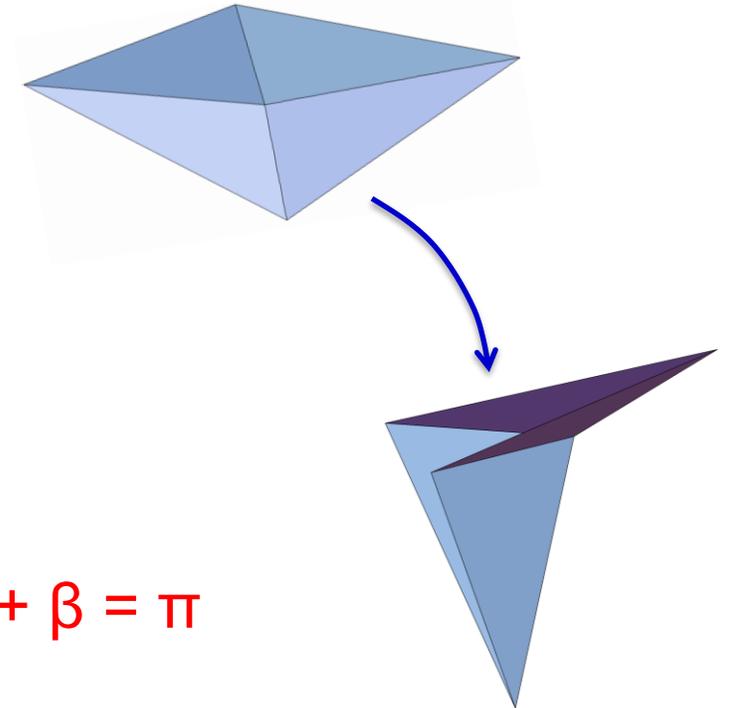
- ❖ If the antennas are placed at positions defined by a group orbit and suitable phases are assigned, strong constructive interference can be achieved in specific (and assignable) directions in the far field.
- ❖ The structure supporting the antennas may have additional functional requirements: reconfigurability, functionality.
- ❖ A natural method: origami design by the group orbit method. Match the groups.

A classic theorem in origami

Theorem. A continuous y satisfies



(two families of solutions)



for A, C in $SO(2)$ and B, D in $SO(2)P$ if and only if $\alpha + \beta = \pi$

In these cases, there exists a continuous folding

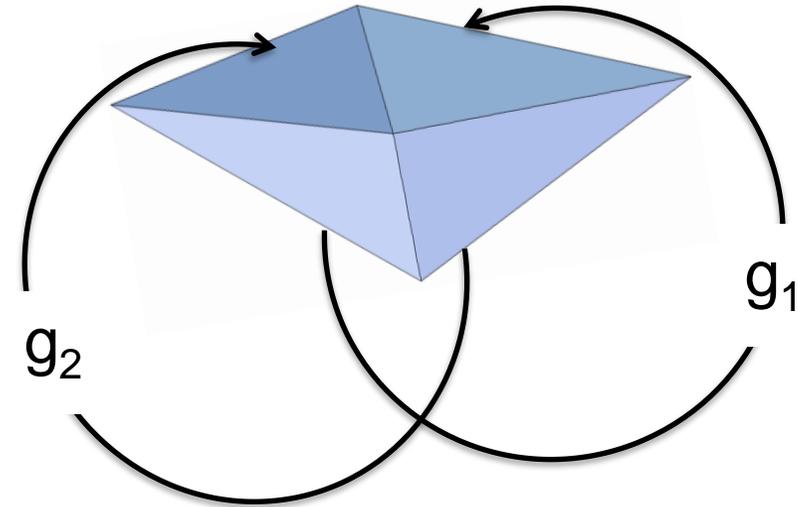
$$\nabla y_\omega : \Omega \rightarrow \mathbb{R}^3, \quad -\pi \leq \omega < \pi \quad \text{where} \quad \nabla y_\omega \in O(2, 3)$$

Where does the group come from?

Two isometries :

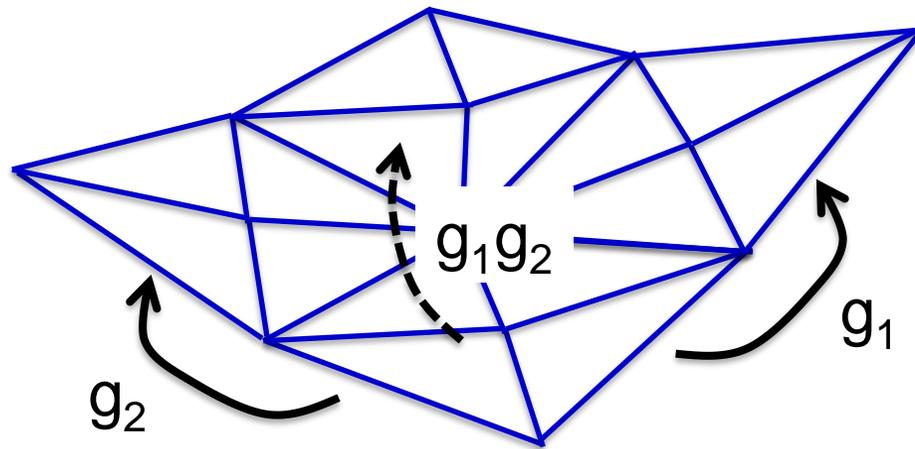
$$g_1 = (\mathbf{Q}_1 | \mathbf{c}_1) \text{ and } g_2 = (\mathbf{Q}_2 | \mathbf{c}_2),$$

$$g_1 g_2 = g_2 g_1$$



$$G = \{g_1^i g_2^j : i, j \in \mathbb{Z}\}$$

Get a compatible
origami structure:



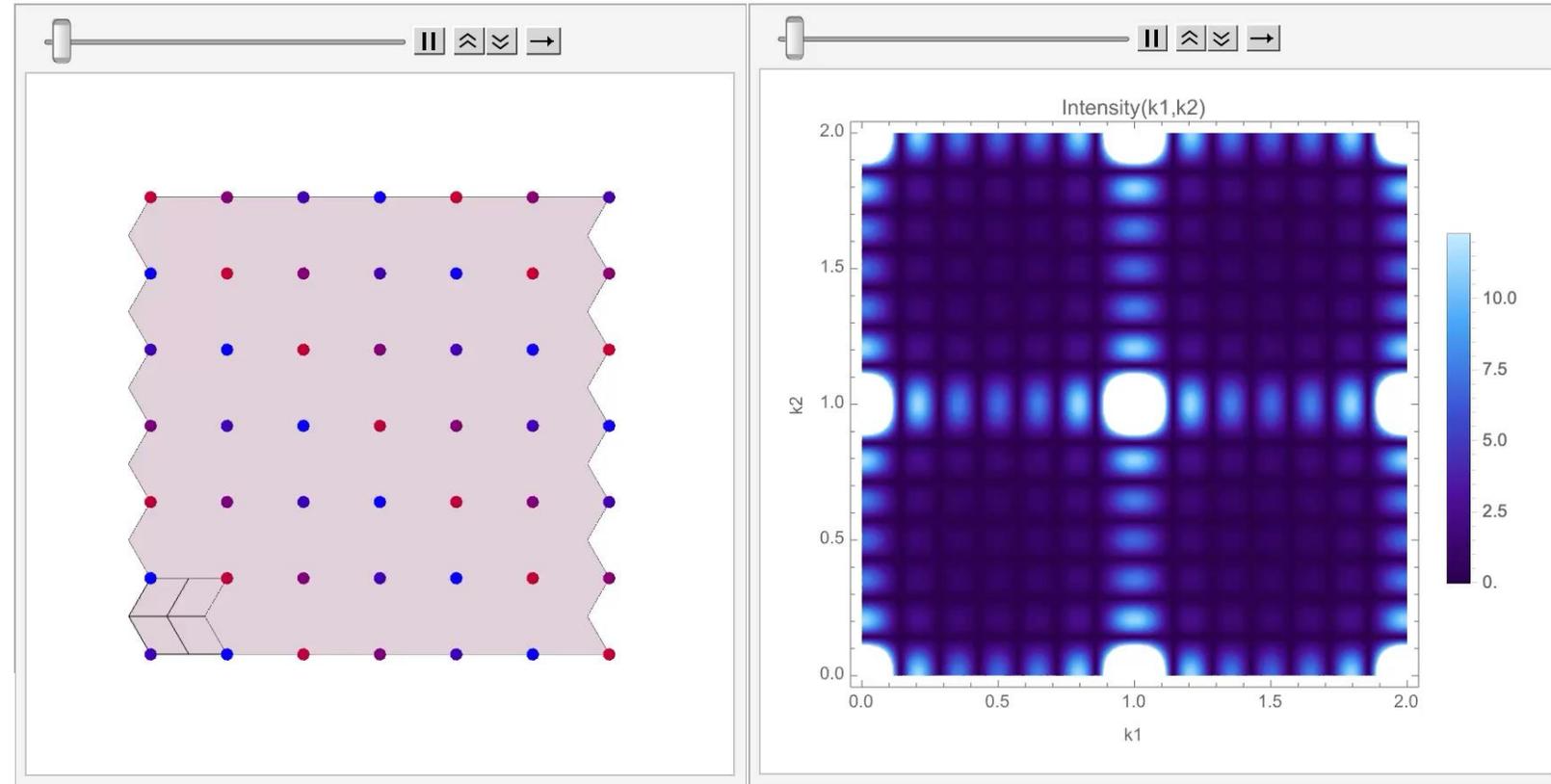
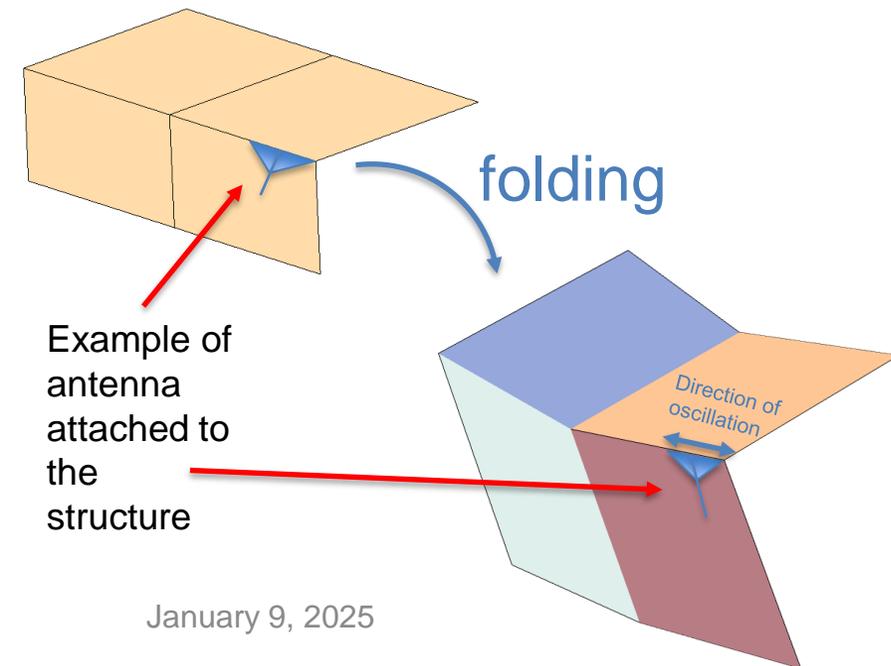
compatibility = commutivity

Application 1 : Monitoring deployable structure

$$\varphi_{lm} = 2\pi (lk_1 + mk_2)$$

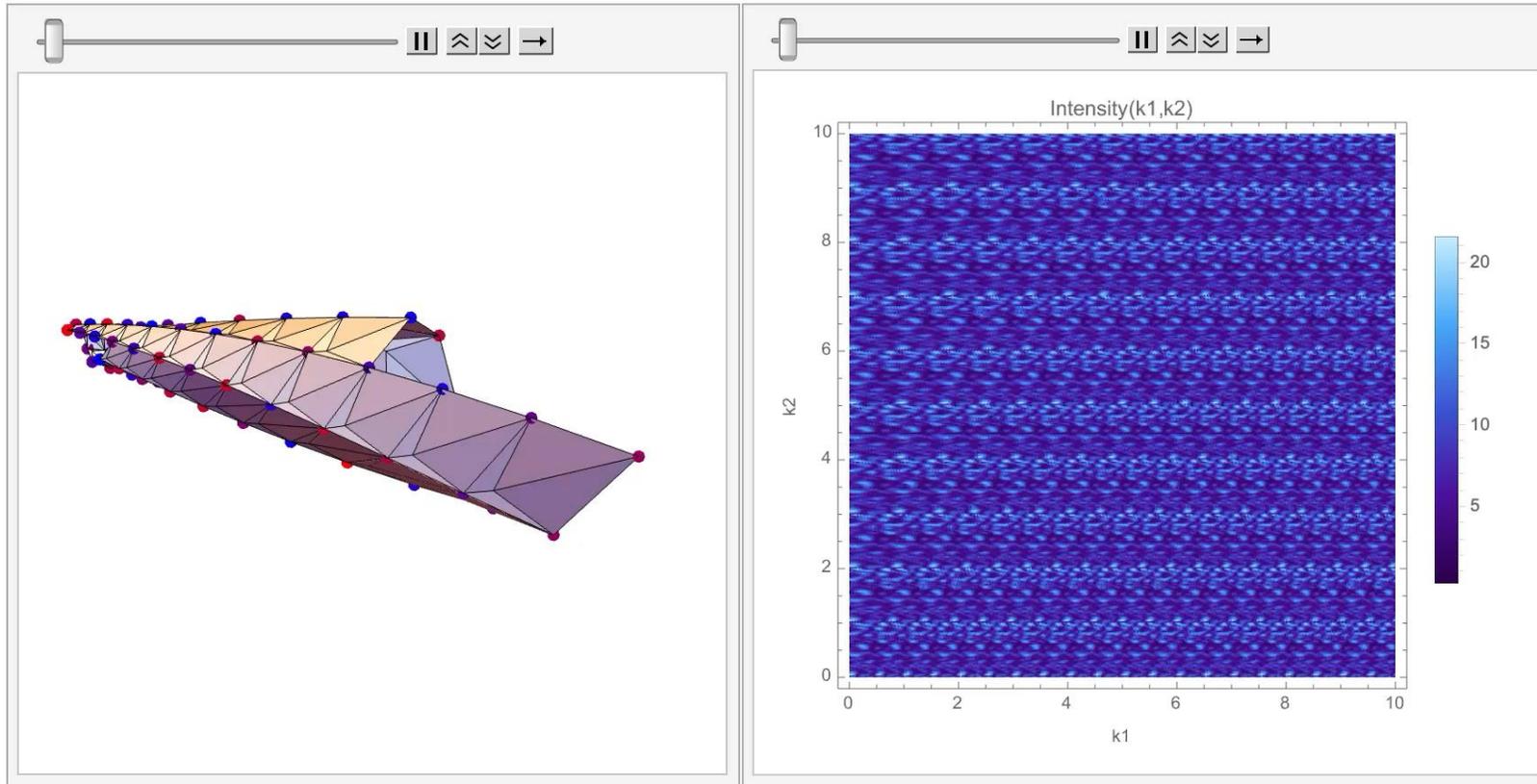
Color of the antenna represents its relative phase

Observation point is along a direction perpendicular to the structure in the far field

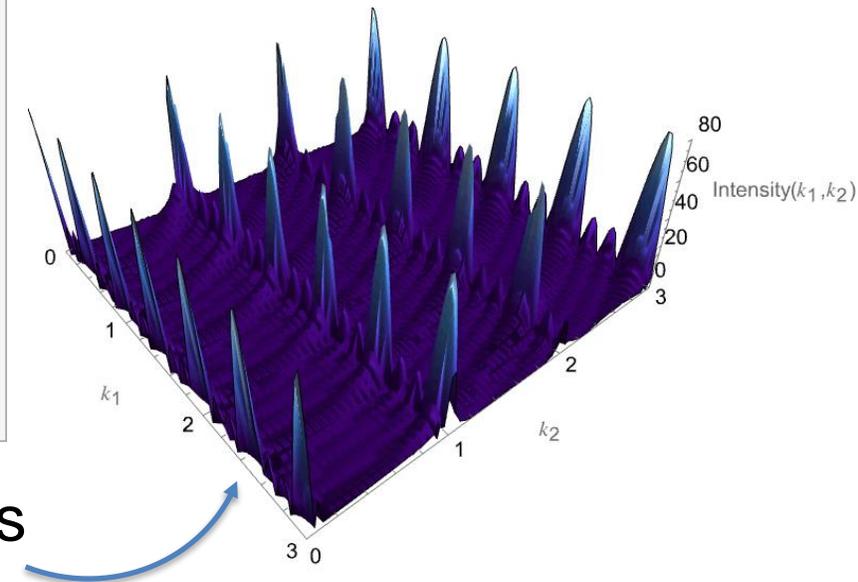


Based on the interference pattern in parameter space, progress of deployment can be determined.

Application 2 : Validating the deployment



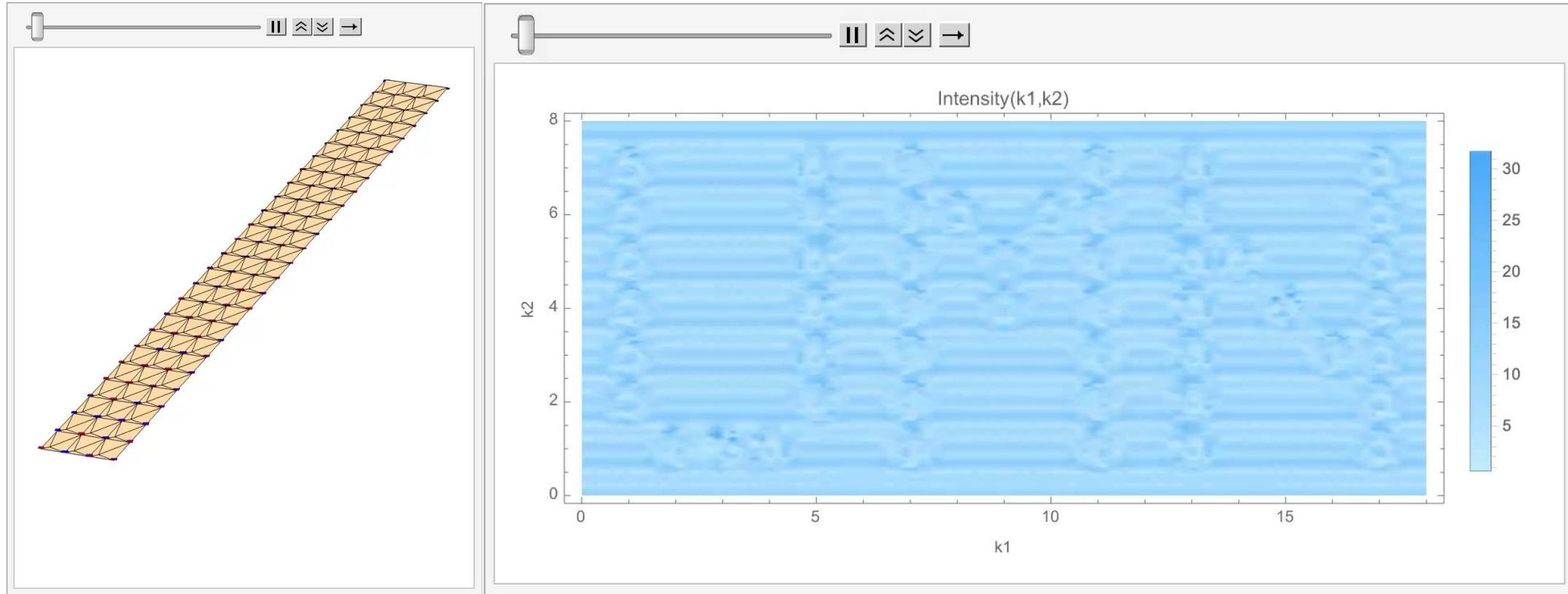
Observation point is along a direction perpendicular to screen in far field



We get the constructive interference when the structure is completely undeformed to a flat state

Application 3 : Display in parameter space

Observation point is along a direction perpendicular to screen in far field



When the structure is completely folded into helical shape, we can see the letters “UMN” within interference pattern in parameter space.

Back to Maxwell's equation, but in free space

$$\mathbf{E}(\mathbf{x}, t) = \frac{1}{2\pi} e^{-i\omega t} \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \mathbf{T}_\omega(k_x, k_y) e^{i(k_x x + k_y y + \gamma z)} dk_x dk_y$$

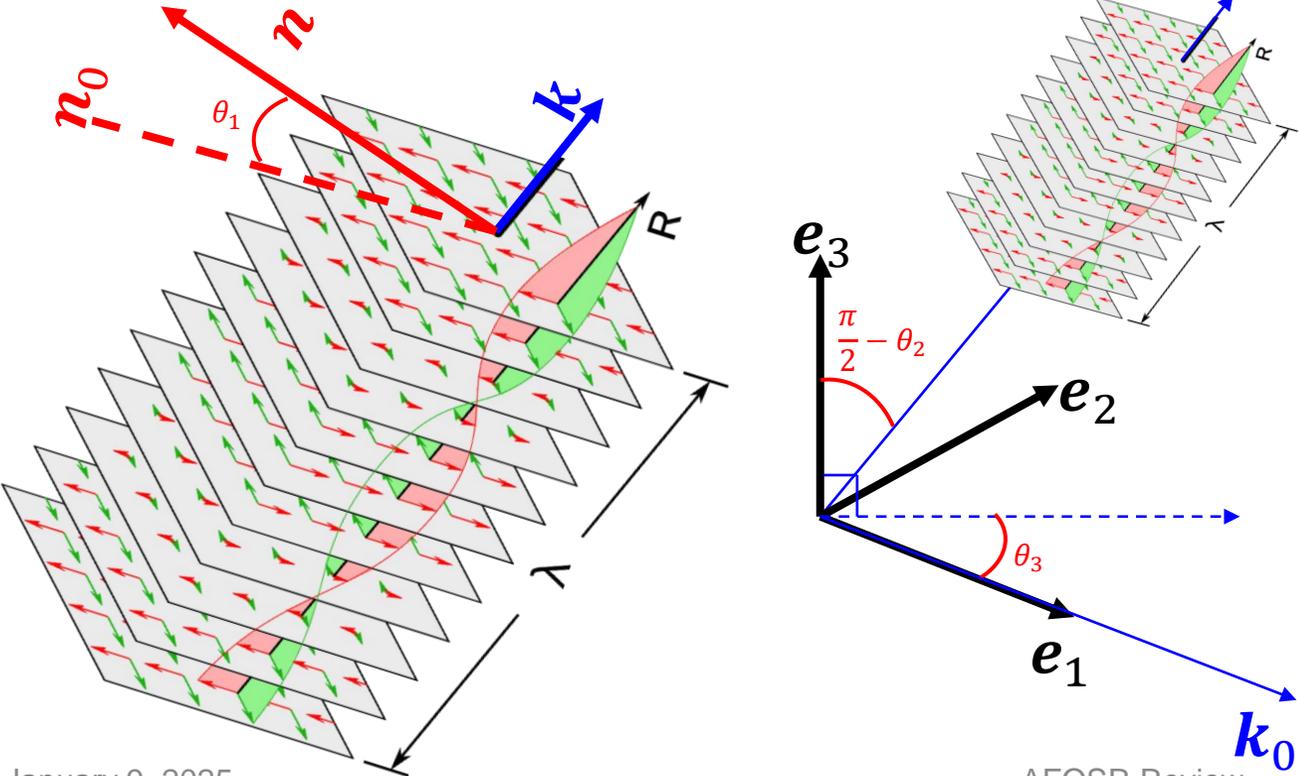
Constraints : $\left\{ \begin{array}{l} \bullet k_x^2 + k_y^2 + \gamma^2 = \omega^2 / c^2 \quad (\text{Monochromatic solution}) \\ \bullet \mathbf{T}_\omega \cdot (k_x, k_y, \gamma) = 0 \end{array} \right.$

Another approach...

$$\mathbf{E}(\mathbf{x}, t) = e^{-i\omega t} \int_{-\pi}^{\pi} \int_{-\pi/2}^{\pi/2} \int_{-\pi}^{\pi} A(\theta_1, \theta_2, \theta_3) \mathbf{R}_3(\theta_3) \mathbf{R}_2(\theta_2) \mathbf{R}_1(\theta_1) \mathbf{n}_0 e^{i(\mathbf{R}_3(\theta_3) \mathbf{R}_2(\theta_2) \mathbf{k}_0 \cdot \mathbf{x})} d\theta_1 d\theta_2 d\theta_3$$

Freedom, we have

$\mathbf{R}_i(\theta_i)$ is rotation about \mathbf{e}_i by θ_i .



General idea :

- Start with $\mathbf{E}_0(\mathbf{x}) = \mathbf{n}_0 e^{i\mathbf{k}_0 \cdot \mathbf{x}}$ such that $|\mathbf{k}_0|^2 = \omega^2/c^2$.
- Rotate this in all possible direction and superimpose

Approximation by trapezoidal rule

$$\mathbf{E}(\mathbf{x}, t) \approx e^{-i\omega t} \sum_l \sum_m \sum_n A_{lmn}(\theta_{1_l}, \theta_{2_m}, \theta_{3_n}) \underbrace{\mathbf{R}_3(\theta_{3_n}) \mathbf{R}_2(\theta_{2_m}) \mathbf{R}_1(\theta_{1_l}) \mathbf{n}_0}_{\text{Maxwell's equations are satisfied at the discrete level as well.}} e^{i(\mathbf{R}_3(\theta_{3_n}) \mathbf{R}_2(\theta_{2_m}) \mathbf{k}_0 \cdot \mathbf{x})}$$

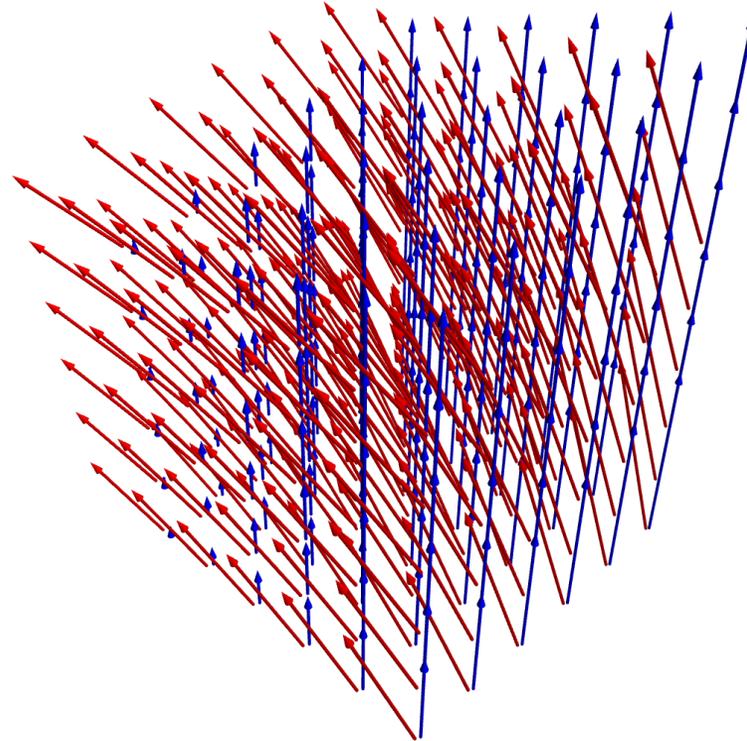
Choose different amplitudes $A_{lmn}(\theta_{1_l}, \theta_{2_m}, \theta_{3_n})$ in approximate formula or $A(\theta_1, \theta_2, \theta_3)$ in exact solution to design different radiations

Maxwell's equations are satisfied at the discrete level as well.

Fast convergence of trapezoidal rule (Illustrative example)

Red – exact solution

Blue – approximate solution

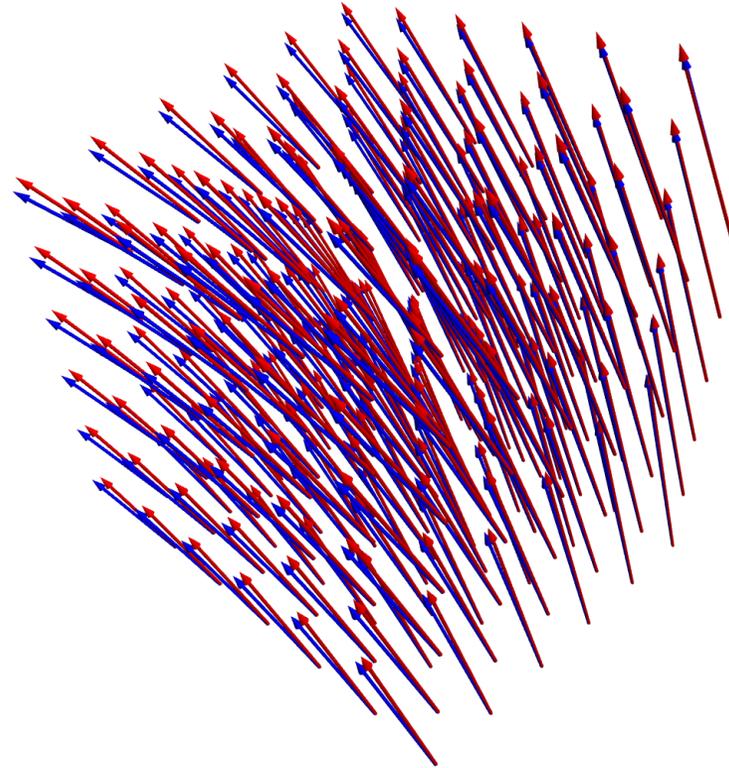


<input type="checkbox"/> Number of terms	16
<input type="checkbox"/> Maximum normalized error	0.32

Fast convergence of trapezoidal rule (Illustrative example)

Red – exact solution

Blue – approximate solution

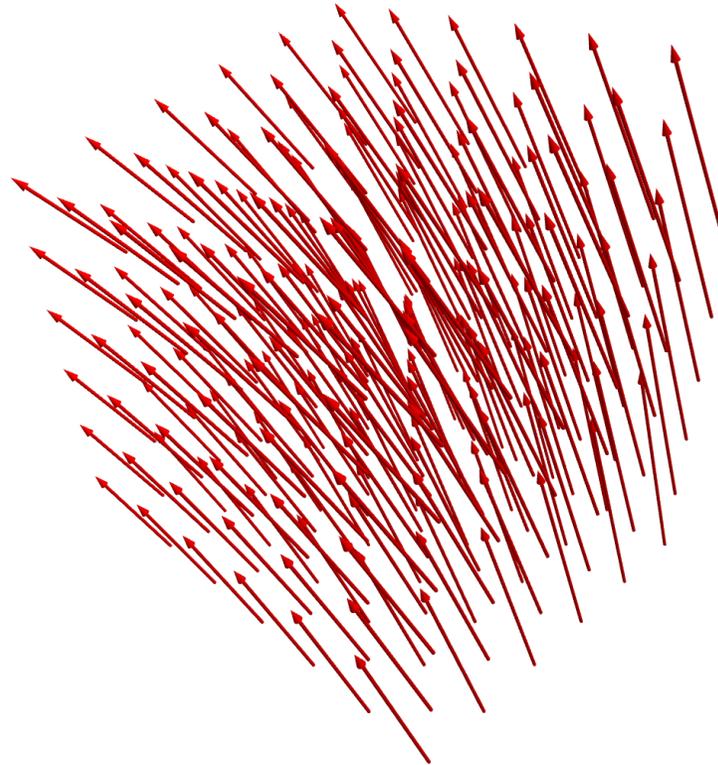


<input type="checkbox"/> Number of terms	36
<input type="checkbox"/> Maximum normalized error	0.11

Fast convergence of trapezoidal rule (Illustrative example)

Red – exact solution

Blue – approximate solution



☐ Number of terms

64

☐ Maximum normalized error

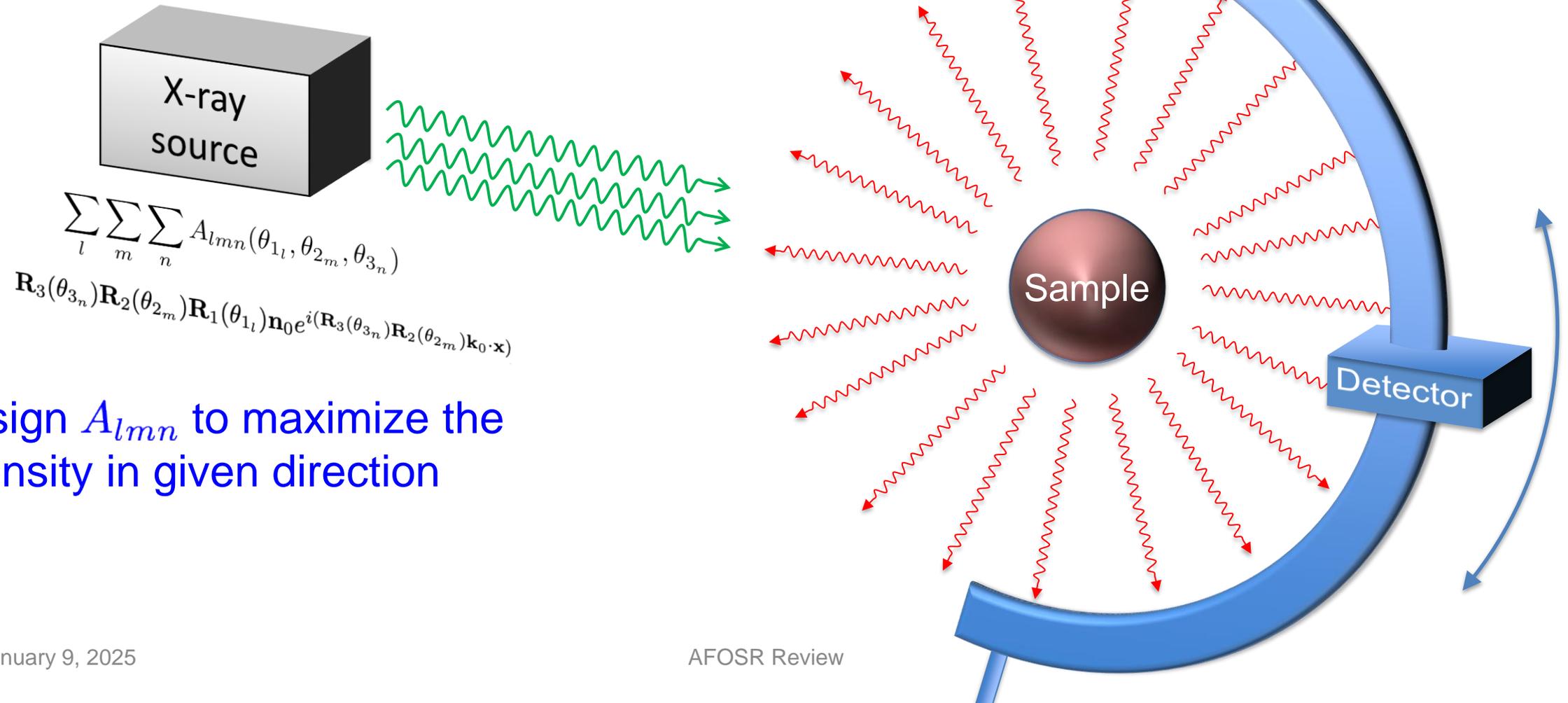
6.26×10^{-6}

Why this solution is interesting?

$$\mathbf{E}(\mathbf{x}, t) \approx e^{-i\omega t} \sum_l \sum_m \sum_n A_{lmn}(\theta_{1_l}, \theta_{2_m}, \theta_{3_n}) \mathbf{R}_3(\theta_{3_n}) \mathbf{R}_2(\theta_{2_m}) \mathbf{R}_1(\theta_{1_l}) \mathbf{n}_0 e^{i(\mathbf{R}_3(\theta_{3_n}) \mathbf{R}_2(\theta_{2_m}) \mathbf{k}_0 \cdot \mathbf{x})}$$

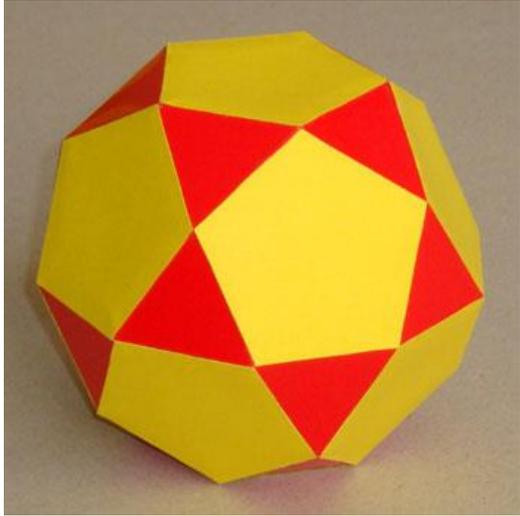
- Exact solution of the time harmonic Maxwell's equations
- Extremely fast convergence of approximation by the Trapezoidal Rule
- The terms in the approximation are synchronized “dipole antennas”
- Time harmonic Maxwell's equations are satisfied at both the discrete level and for the limit

Schematic for structure determination experiments



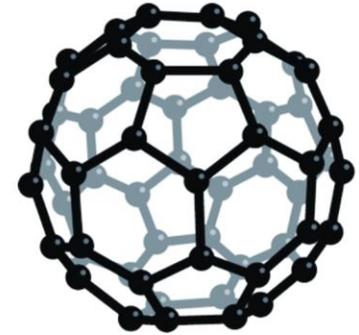
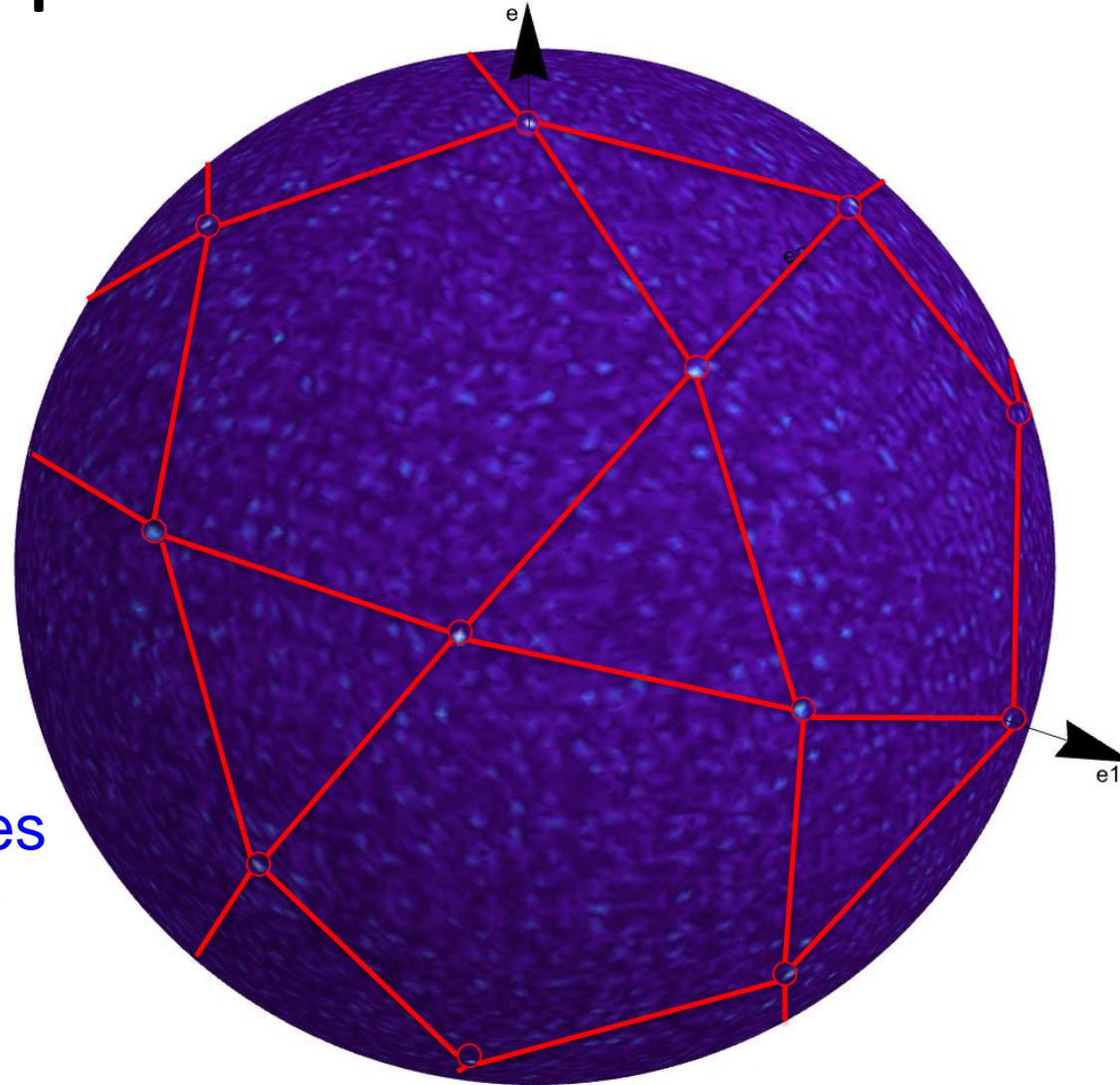
Design A_{lmn} to maximize the intensity in given direction

Intensity pattern with molecules like C_{60}



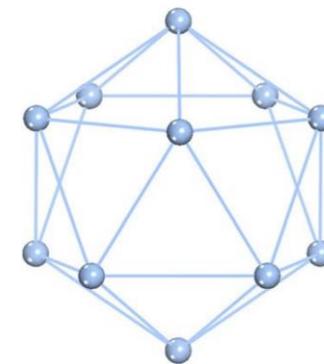
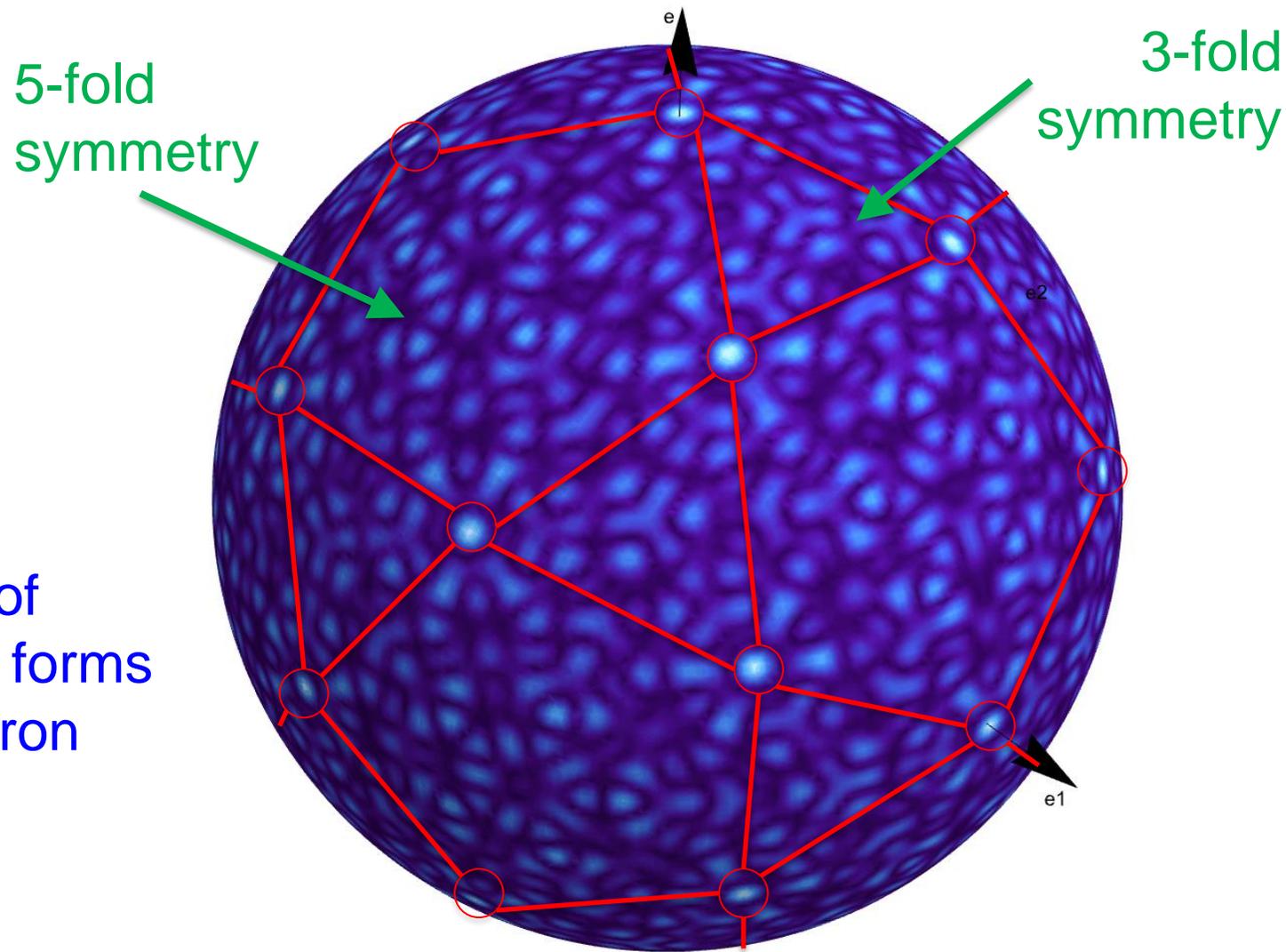
icosidodecahedron

Vertices of peak intensities
forms icosidodecahedron



C_{60} molecule

Intensity pattern with molecule like Ag_{12}

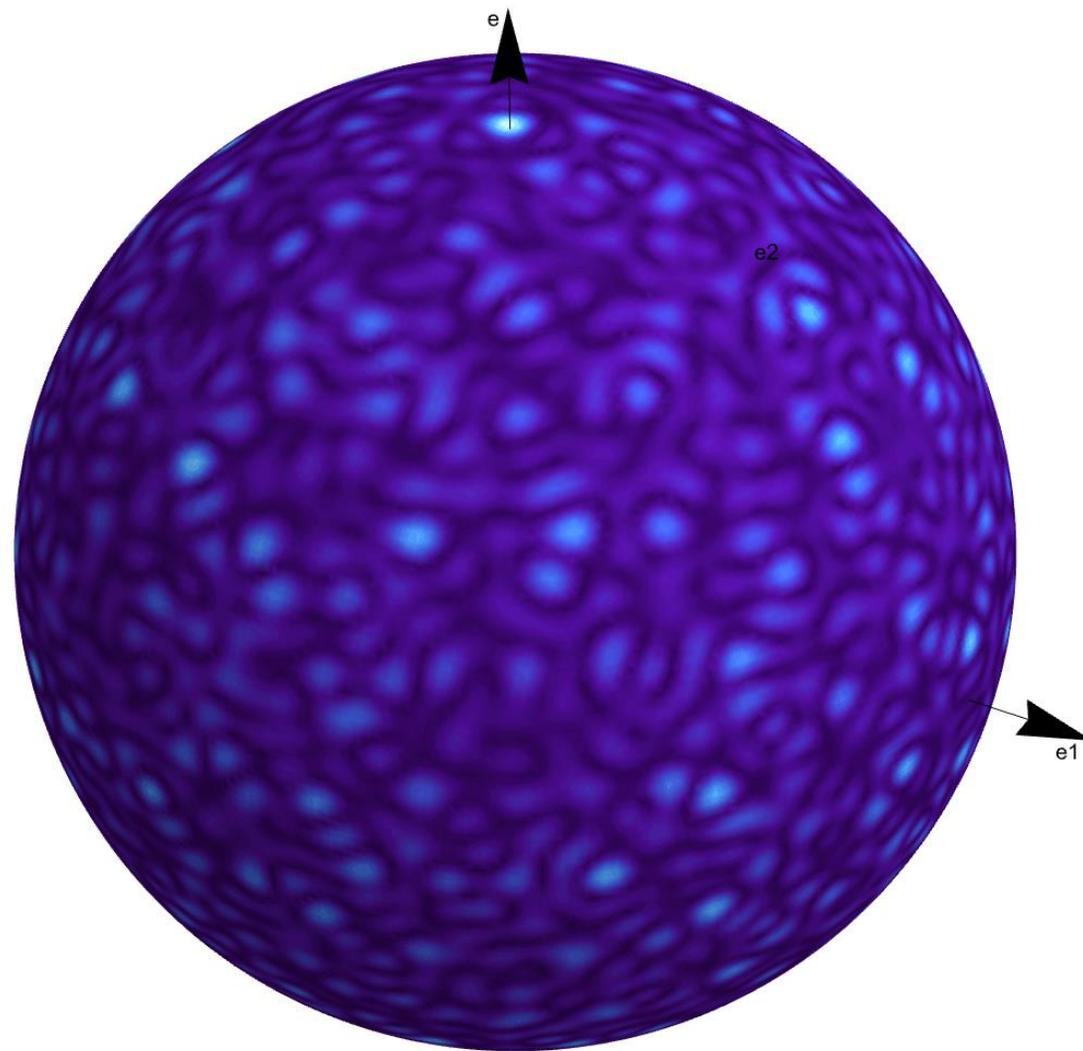


Ag_{12} molecule

Again, vertices of peak intensities forms icosidodecahedron

Intensity pattern with molecule like Au_{30}

Need to solve the
inverse problems for
these materials



Thank you

Questions?

Applications in radar technology

